

two CP states. Turning a specific p-i-n diode on and other one off, the antenna can operate with RHCP and by interchanging the states of the p-i-n diodes, LHCP radiation can be obtained. A commercial electromagnetic solver.

III. MECHANISM OF CIRCULAR POLARIZATION(MSD)

MCD observed from +Z direction keeping one of the PIN-diodes on and other one off at the f_c of 1.75GHz at two phases: $\omega t = 0^0, 90^0$ are shown in Fig. 2. When the SW1 is in on condition and SW2 in off condition, the MCD is in anti-clockwise direction. So, in +Z direction obtained radiation would be RHCP, and that in -Z direction would be LHCP.

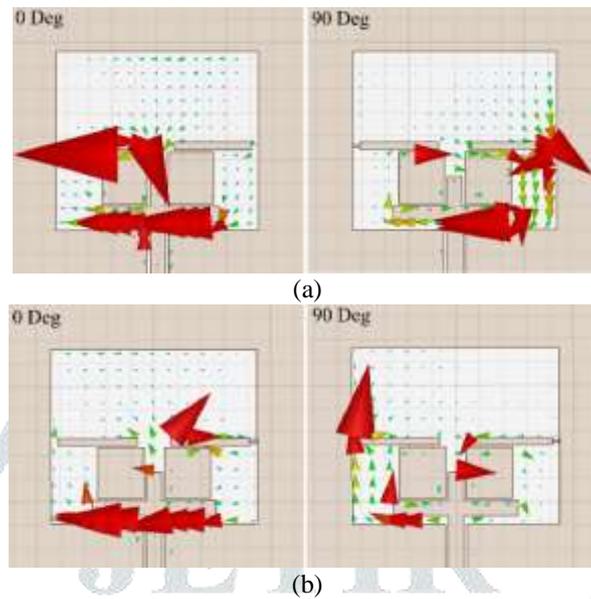


Fig. 2. Magnetic current distribution at 1.75GHz (a) diode 1 on diode 2 off and (b) diode 1 off diode 2 on; for proposed antenna at two different time instants: $\omega t = 0^0$ and 90^0

Similarly, Fig. 2(b) shows the MCD for SW1 off and SW2 on at the f_c of 1.75GHz at two phases: $\omega t = 0^0, 90^0$ observed from the +Z direction. In +Z direction obtained radiation would be LHCP, and that in -Z direction would be RHCP because rotation of magnetic current is clockwise in this case.

IV. ANTENNA DESIGN AND ANALYSIS

Sequential improvement is shown by four antennas of Fig. 3. Taking a quarter wavelength (with respect to the first resonating frequency of the slot) long straight feed as shown in Ant 1 of Fig.3, the slot can generate linear polarized wave. Guided wavelength is considered as $\lambda_g = \lambda_0 / \sqrt{\epsilon_e}$, where λ_0 is free space wavelength and $\epsilon_e = (\epsilon_r + 1)/2$. Then, placing two metallic strips connected to the center of the two opposite sides of the slot edges by p-i-n diodes and turning one of them on as shown in Ant 2, E- field of the slot is perturbed and two orthogonal E- field components having same magnitudes and quadrature phases are excited as the strip connected to the switched on p-i-n diode acts as grounded stub. Here, feed length is adjusted to get best possible impedance matching. Then, placement of pair of rectangle shaped parasitic patches as in Ant 3 improves axial ratio bandwidth. It reduces also the operating frequency of CP operation because due to the patches the paths of the magnetic current become zigzag and thereby lengthy. In Ant 1 to Ant 3, except length of the feed lines, other dimensions of the antennas are kept to the optimized values. Finally placement of horizontal branches of the feed line as shown in Ant 4 increases the AR bandwidth further, reduces operating frequency of CP operation more, and improves the impedance matching a lot.

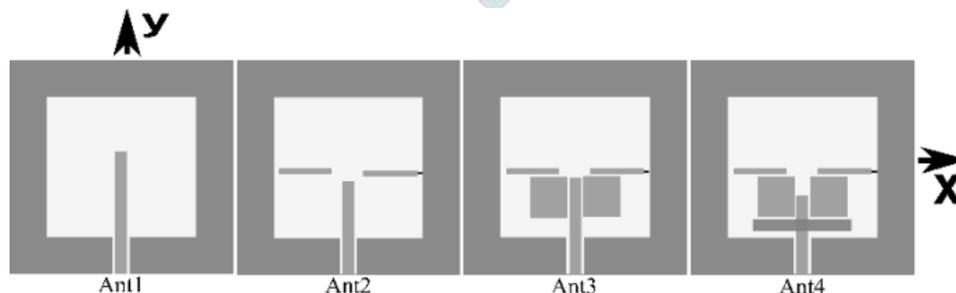


Fig. 3. Steps of improvement of the proposed antenna.

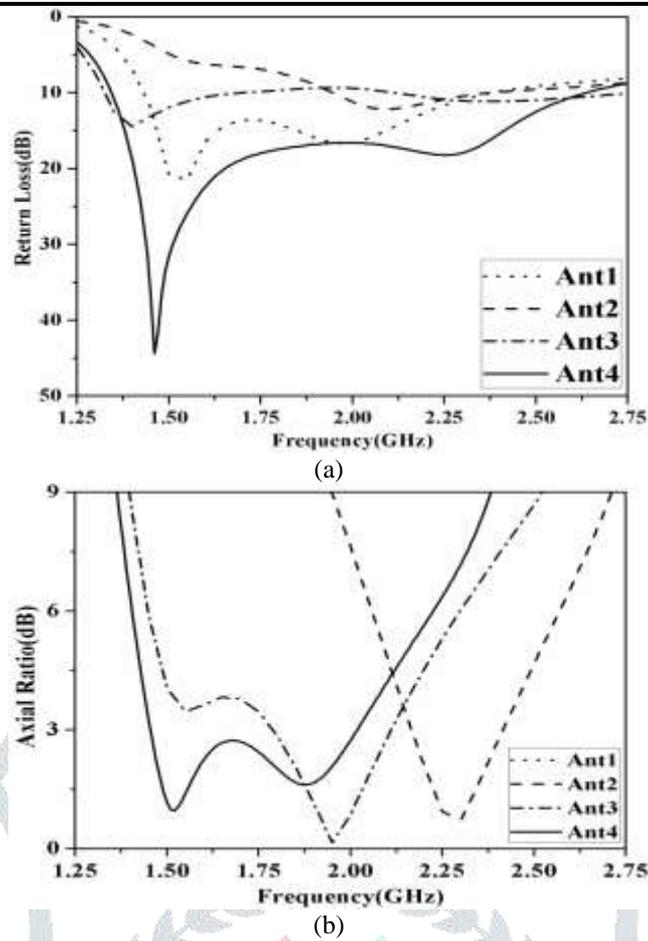


Fig.4. Simulated diagrams of (a) the return loss (dB) and (b) the axial ratio for Ant 1-4.

To obtain good performance, optimizing different parameters one by one and observing their effects on the impedance, and axial ratio bandwidth of the antenna, the different dimensions of the proposed antenna have been selected. The parameters of the feed line have more effect on impedance matching as shown in Fig. 5(a) & 6(a). However, they have also some effect on AR as shown in Fig. 5(b) & 6(b) because its dimension perturbs the MCD within the dielectric part of antenna. It is observed that max ARBW is achieved when $L_1=21.3$ mm, and $L_2=26.4$ mm.

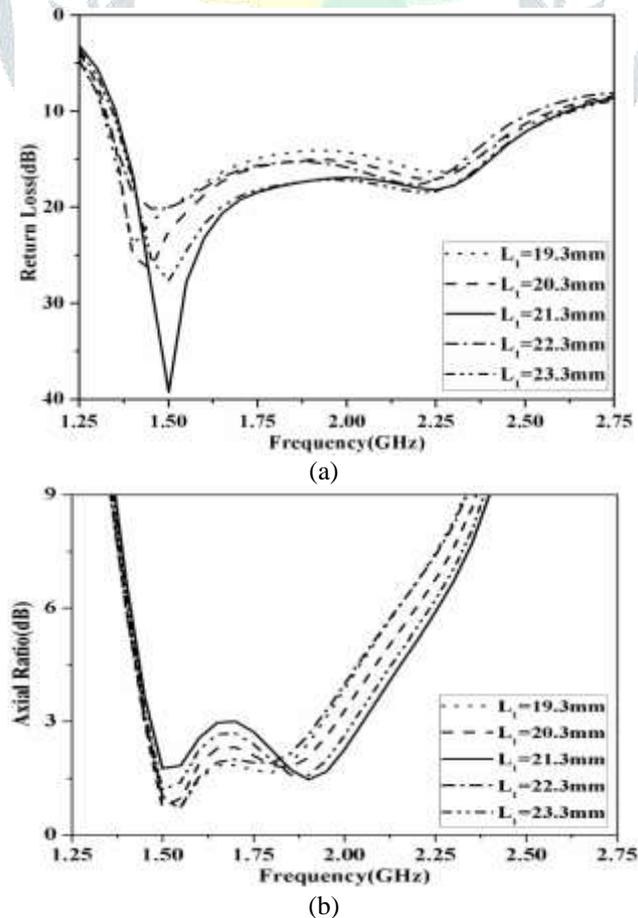
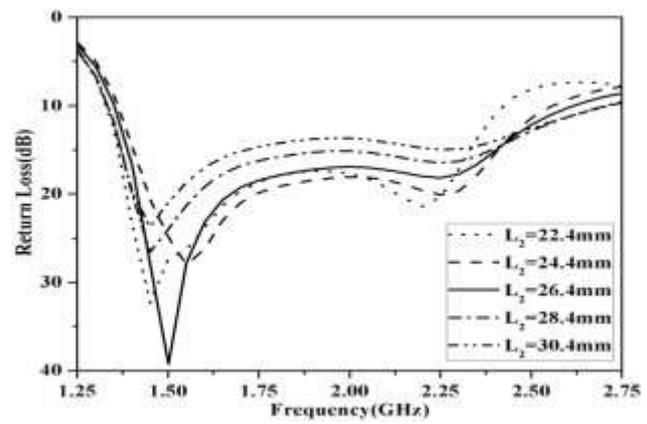
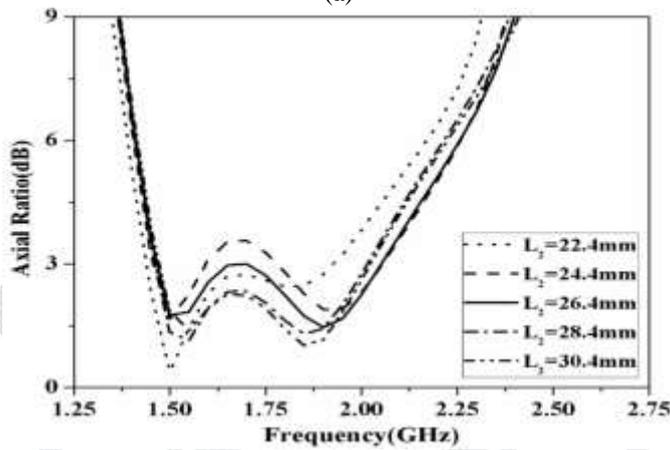


Fig. 5. Effect of parameter L_1 on antenna parameters: (a) the return loss (b) the axial ratio.



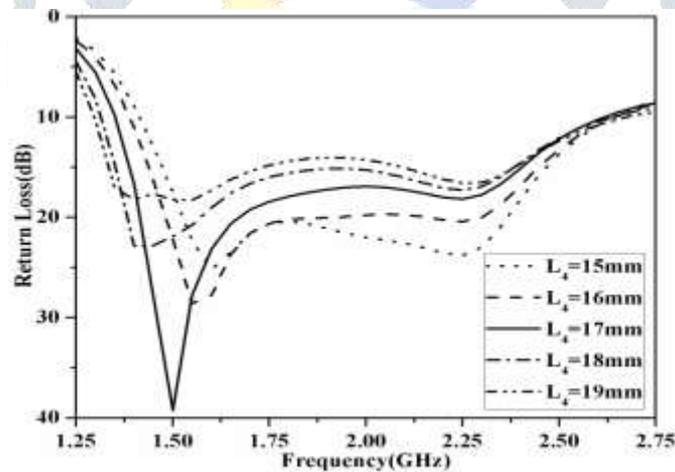
(a)



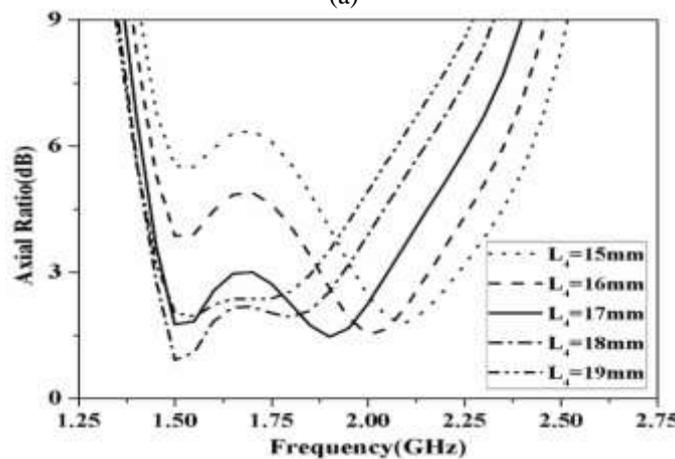
(b)

Fig. 6. Effect of parameter L_2 on antenna parameters: (a) the return loss (b) the axial ratio.

Figure 7(a) and (b) show the effect of length L_4 of strips on antenna performance. It has been seen that length of strips has more effect on AR and AR bandwidth is maximum when $L_4=17$ mm.



(a)



(b)

Fig. 7. Effect of the Length of strips on antenna parameters: (a) the return loss (b) the axial ratio.

Figure 8(a) and (b) show the effect of parasitic patches on antenna performance. Variations of the size of the patches are done with respect to their center points. It has been observed that patches have effect on AR because they perturb the magnetic current distribution within the slot. They have effect on impedance matching also as they are very close to feed lines. Best performance of the antenna in terms of both impedance matching and AR is observed when $L_3=11$ mm, and $W_3=10$ mm.

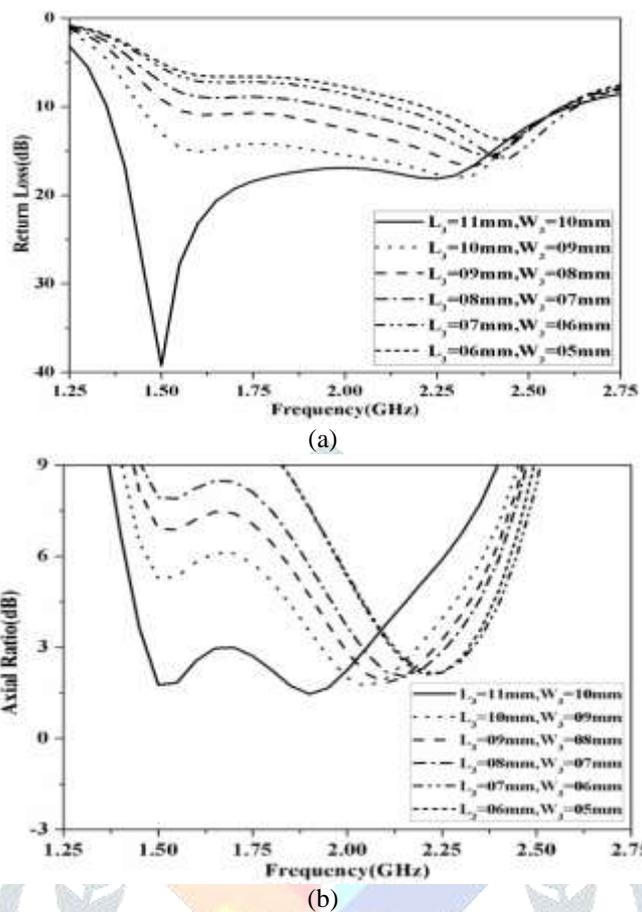


Fig. 8. Effect of the length & width of parasitic elements: (a) the return loss (b) the axial ratio.

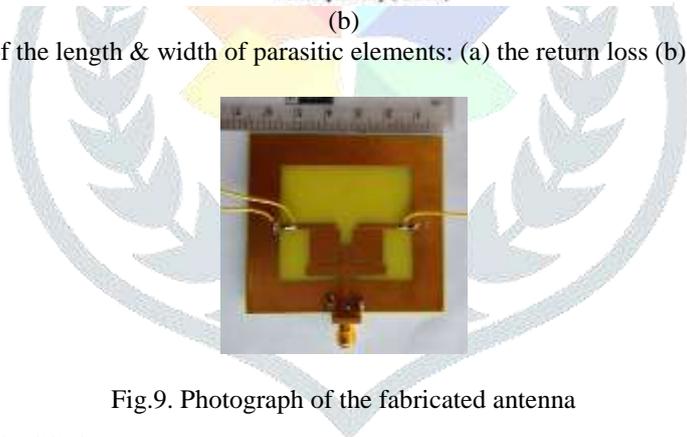


Fig.9. Photograph of the fabricated antenna

V. RESULT AND DISCUSSION

The fabricated antenna is shown in Fig. 9. When the pin-diode SW1 is in on condition and SW2 is in off condition, RHCP is obtained. On the other hand, when SW2 is on condition and SW1 in off condition, LHCP is obtained in +Z direction.

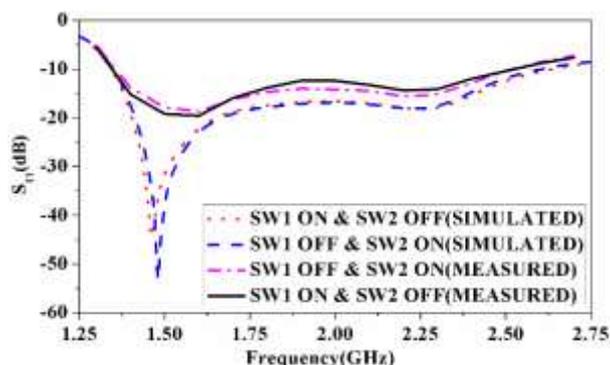


Fig. 10 Simulated & Measured S11(dB) of the proposed geometry of antenna.

The result of measured and simulated reflection coefficient are shown in Fig.10. The return-loss bandwidth of the fabricated antenna has the frequencies from 1.5GHz to 2.38GHz.

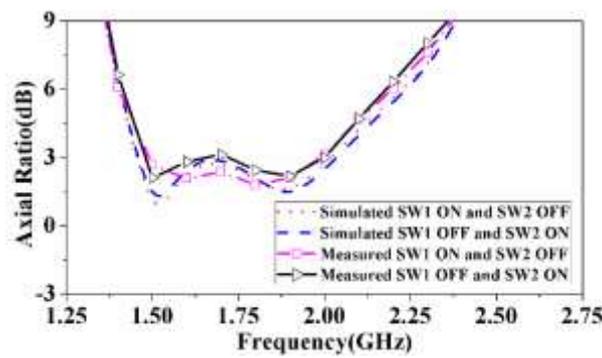


Fig. 11. Measured and simulated AR curves of the proposed antenna.

Figure 11 illustrates the measured and simulated axial ratio. The measured CP bandwidth determined by the 3-dB AR is about 420 MHz or 23.21% (1.6GHz–2.02GHz).

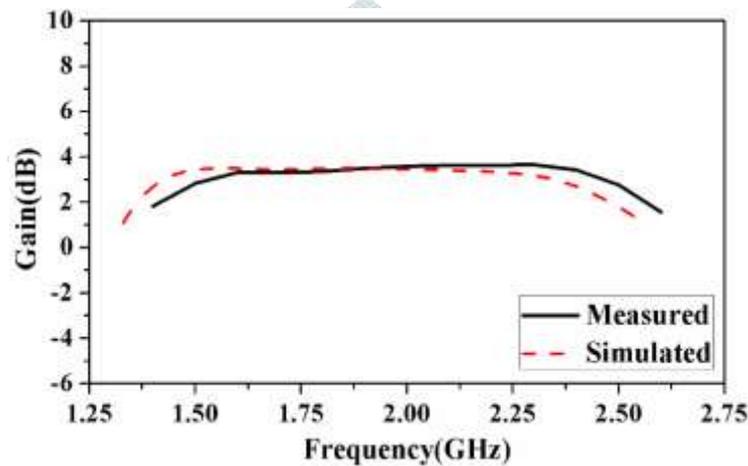


Fig. 12. Measured and simulated gain in +Z direction of the proposed antenna.

Figure 12 shows the gain of the final antenna. The gain deviates from 1.5dB to 3dB in ARBW. Figures 13 show the radiation patterns of the antenna at 1.75GHz. Cross polarization level is about 18 dB lower than co-polarization in the direction of maximum radiation (+z). When the pin-diode SW2 is in on condition and SW1 in off condition, the polarization is reversed. There is some mismatches between the results due to non-perfect fabrication.

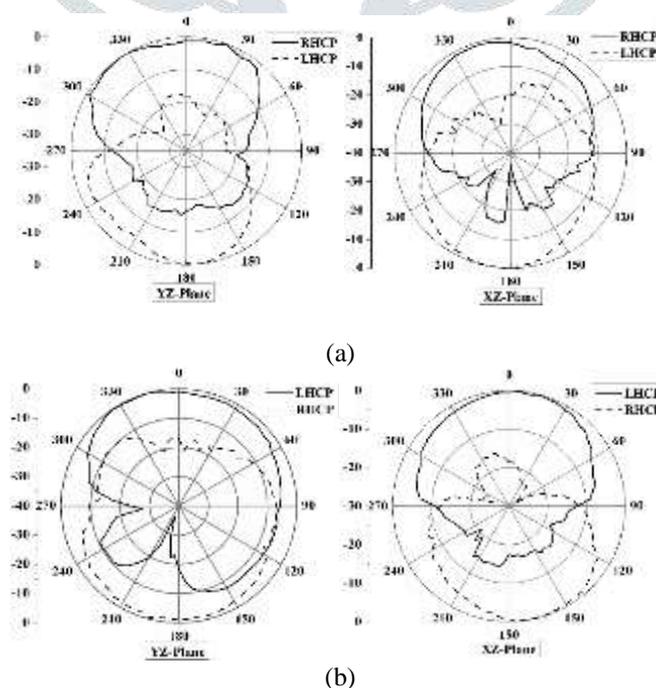


Fig. 13. Measured radiation patterns at 1.75GHz of the proposed antenna (a) SW1 on & SW2 off and (b) SW1 off & SW2 on.

VI. CONCLUSION

A broadband polarization diversity antenna is presented with a cross shaped CPW feed, a pair of patches and a pair of stubs. The final antenna has wide impedance bandwidth of 45.36% (1.5GHz-2.38GHz) and AR bandwidth of 23.21% (1.6GHz-2.02GHz). The proposed antenna is useful in Global Positioning System (GPS; 1.575GHz) and Digital Communication System (DCS; 1.71-1.88GHz).

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