



FPGA Implementation of Multi Error Correction and Detection for IOT Applications

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Abstract : Wireless communication network must be able to transfer data with high accuracy and reliability for various application. So to retain the more reliability of these networks, errors must be detected and corrected with great efficiency. There are number of error detecting and correcting codes available but most of them confront lot of challenges such as unreliable wireless links, the broadcast nature of wireless transmissions, interference, frequent topology changes, and other effects of wireless channels. This work proposed an implementation of multi error correction and detection using verilog code on xilinx 14.7 software. Proposed MEC-MED circuit design for high speed and less complexity and reduced area logic circuit. The conventional SEC-SED logic is modified to extend error correction and detection and improve latency, area, power, throughput and frequency. Simulated results shows that proposed MEC-MED VLSI architecture gives significant improved results than conventional SEC-SED.

IndexTerms – SEC, SED, MEC, MED, IOT, WSN, Verilog, VLSI, Xilinx.

I. INTRODUCTION

All error-detection and correction schemes add some redundancy (i.e., some extra data) to a message, which receivers can use to check consistency of the delivered message, and to recover data that has been determined to be corrupted. Error-detection and correction schemes can be either systematic or non-systematic. In a systematic scheme, the transmitter sends the original data, and attaches a fixed number of check bits (or parity data), which are derived from the data bits by some deterministic algorithm. If only error detection is required, a receiver can simply apply the same algorithm to the received data bits and compare its output with the received check bits; if the values do not match, an error has occurred at some point during the transmission. In a system that uses a non-systematic code, the original message is transformed into an encoded message carrying the same information and that has at least as many bits as the original message.

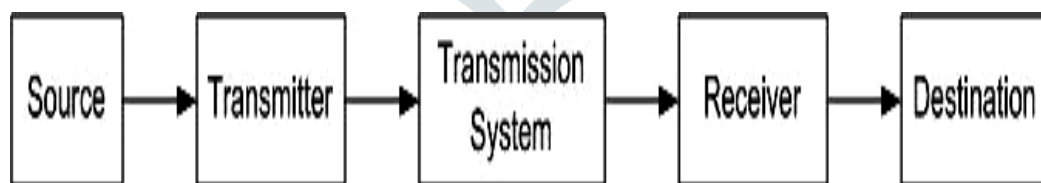


Figure 1.1: Basic Communication

Good error control performance requires the scheme to be selected based on the characteristics of the communication channel. Common channel models include memoryless models where errors occur randomly and with a certain probability, and dynamic models where errors occur primarily in bursts. Consequently, error-detecting and correcting codes can be generally distinguished between random-error-detecting/correcting and burst-error-detecting/correcting. Some codes can also be suitable for a mixture of random errors and burst errors.

Convolutional codes are processed on a bit-by-bit basis. They are particularly suitable for implementation in hardware, and the Viterbi decoder allows optimal decoding. Block codes are processed on a block-by-block basis. Early examples of block codes are repetition codes, Hamming codes and multidimensional parity-check codes. They were followed by a number of efficient codes, Reed-Solomon codes being the most notable due to their current widespread use. Turbo codes and low-density parity-check codes (LDPC) are relatively new constructions that can provide almost optimal efficiency. Shannon's theorem is an important theorem in forward error correction, and describes the maximum information rate at which reliable communication is possible over a channel that has a certain error probability or signal-to-noise ratio (SNR). This strict upper limit is expressed in terms of the channel

capacity. More specifically, the theorem says that there exist codes such that with increasing encoding length the probability of error on a discrete memoryless channel can be made arbitrarily small, provided that the code rate is smaller than the channel capacity. The code rate is defined as the fraction k/n of k source symbols and n encoded symbols.

The actual maximum code rate allowed depends on the error-correcting code used, and may be lower. This is because Shannon's proof was only of existential nature, and did not show how to construct codes which are both optimal and have efficient encoding and decoding algorithms.

II. PROPOSED METHODOLOGY

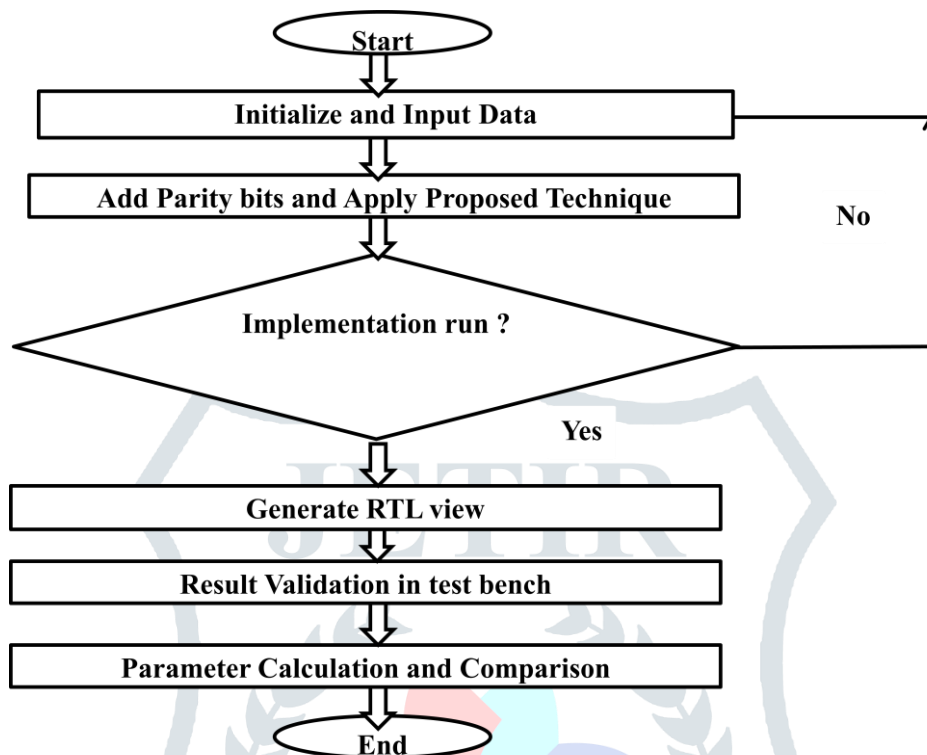


Figure 2: Flow Chart

Modified Single Error Correction - Double Error Detection

The H-matrix for proposed multi error correction- multi error detection (MEC-MED) codes have been constructed using following basic conditions:

1. All the columns are nonzero and distinct.
2. All data columns must have weight (w) three.
3. The XOR sum of any two columns should not be equal to any of the individual column.
4. The XOR sum of any two adjacent columns must be distinct and nonzero. The condition 1 is necessary for the single error correction (MEC) process. Double error detection property is confirmed by conditions 1, 2, and 3.

The procedure to generate the proposed H-matrix for MEC-MED codes is as follows:

Step 1: Initializing the H-matrix with $(n-k)$ number of rows and (n) numbers of columns.

Step 2: The H-matrix consists of $(n-k)$ numbers of parity columns having identity property and k numbers of data columns.

Step 3: Perform the modulo-2 operation between first parity column (p_1) and last data column (d_8).

Step 4: Perform the modulo-2 operation between last data column (d_8) and its previous data column (d_7).

Step 5: Continuing modulo-2 operation till second data column (d_2) and first data column (d_1) is performed.

Step 6: Place with '1' in data column's position which obtain from modulo-2 operations and remaining position with '0'.

III. RESULT AND ANALYSIS

The implementation and simulation of the proposed algorithm is done over Xilinx 14.7. The behavioral modeling style and Isim simulator is adopted for simulation. RTL and synthesis results are also generated.

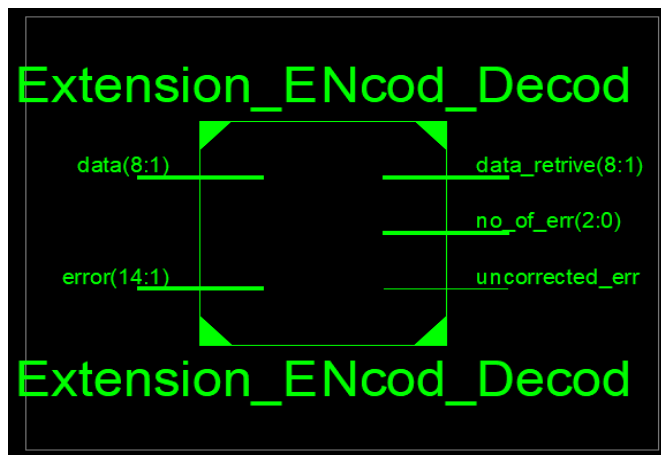


Figure 3: Top level module

Figure 3 is showing the top level module. Here the input is 8bit and apply the error upto 15 bit. The output side of block is 8 bit data retrieval (Recovered original data), number of error is 2 and 1 bit uncorrected error status.

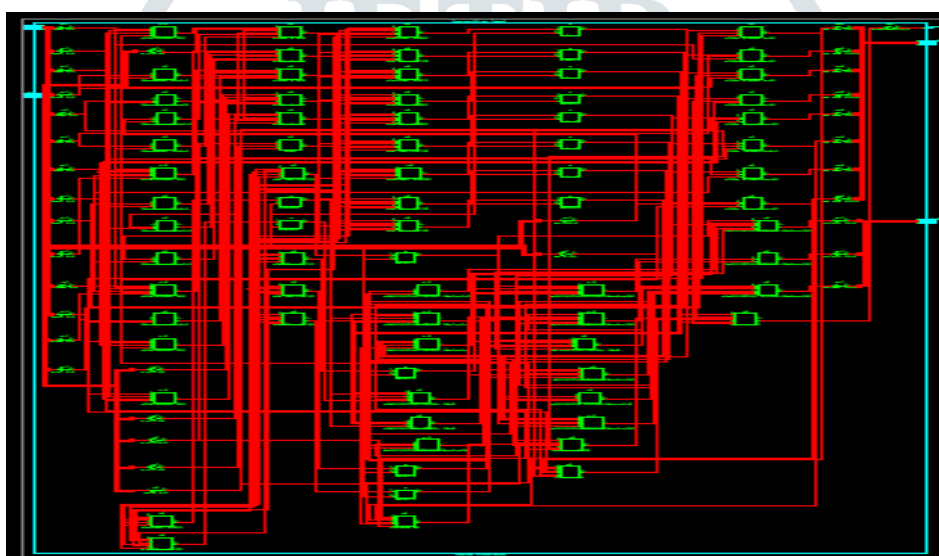


Figure 4: Technological RTL View

Figure 4 is providing the complete technological RTL view of proposed circuits.

Table 1: Device utilization summary

Device Utilization Summary (estimated values)			
Logic Utilization	Used	Available	Utilization
Number of Slice Registers	14	408000	0%
Number of Slice LUTs	61	204000	0%
Number of fully used LUT-FF pairs	14	61	22%
Number of bonded IOBs	34	600	5%

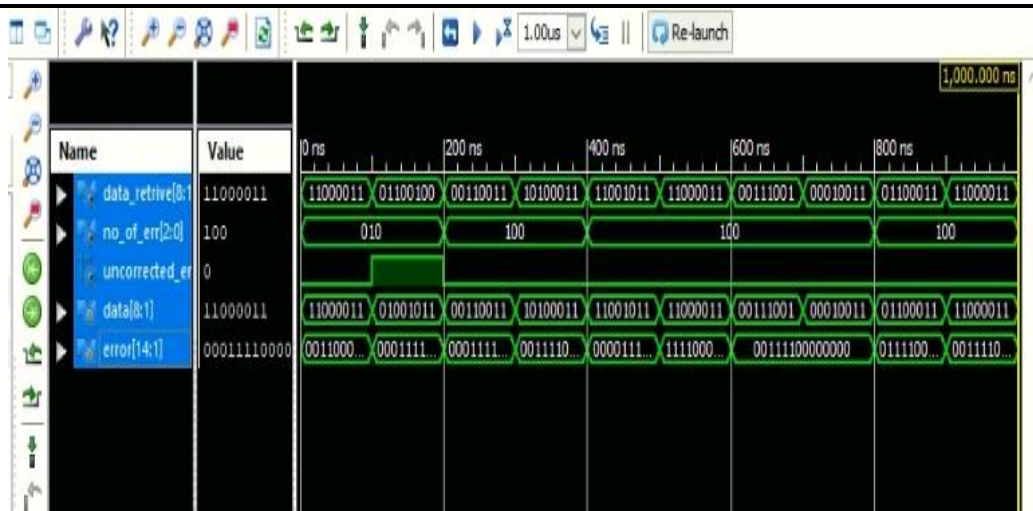


Figure 5: Data input with no error

Figure 5 is providing different input bits and out bits according the error and no error condition. Unexpected error shows bit 1 and signal shows at high conditions.

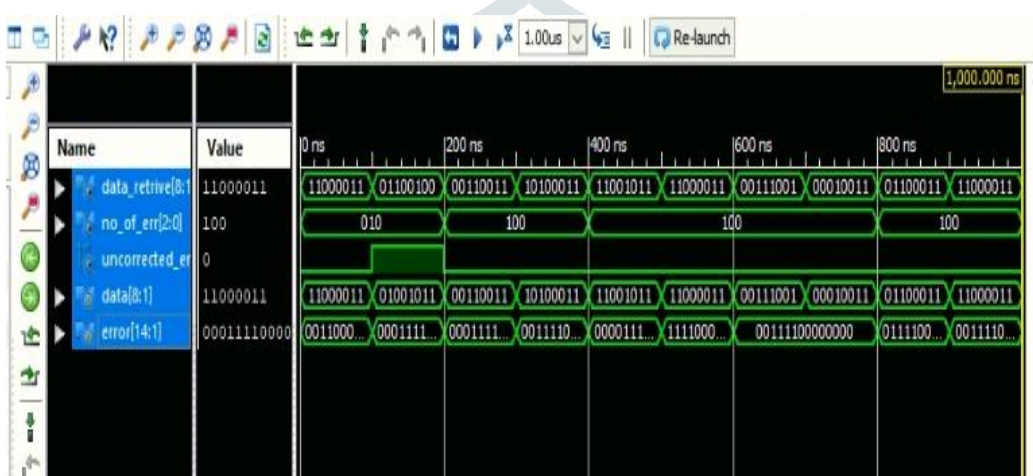


Figure 6: Error Transition zero

Figure 6 is providing different input bits and out bits according the error and no error condition. Error transition is zero in this case.

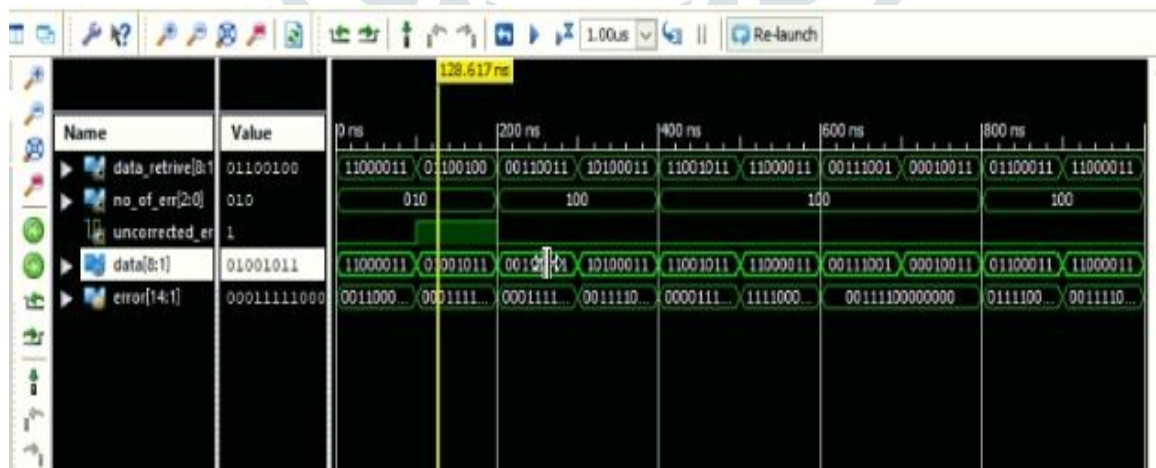


Figure 7: Error Transition from zero to one

Figure 7 is providing different input bits and out bits according the error and no error condition. Error transition is zero to one for a case, for others it's in zero.

Table 2: Simulation Parameters (Virtex 5)

Sr. No	Parameter	Value
1	Area	6.75%
2	Delay	2.538ns
3	Frequency	394MHz
4	Memory	4624652 kilobytes
5	Completion Time	20.00 Secs
6	Throughput	3.15 GHz
7	Power	450 μ W
8	PDP	113.85 fJ

In table 2, simulation parameters are showing which is taken during the execution of verilog script.

Table 3: Comparison chart of proposed work with previous work

Sr. No	Parameters	Previous Work	Proposed Work
1	No of Error	Single	Multi
2	Area	2826.59 μm^2	675 μm^2
3	Delay	434.4 ps	2.538 ns
4	Power	619.47 μ W	450 μ W
5	PDP	269.10 fJ	113.85 fJ

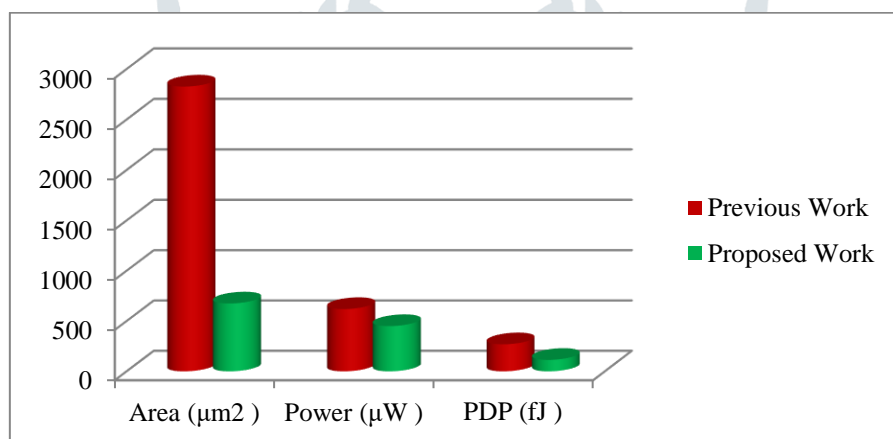


Figure 8: Comparison Graph

Figure 8 is shows the comparison graph of area, power and PDP of proposed and the previous work. It is clear from the comparison table and graphs. It can be say that the proposed work gives the significant better results than the previous work.

IV. CONCLUSION

This work proposed an implementation of multi error correction and detection using verilog code on xilinx 14.7 software. Proposed MEC-MED circuit design for high speed and less complexity and reduced area logic circuit. The conventional SEC-SED logic is modified to extend error correction and detection and improve latency, area, power, throughput and frequency. The conditions for our simulations are: ISim simulator at vertex-v family. The comparison of proposed and previous work is done in terms of the calculated parameters. The proposed work utilized the 675 μm^2 area while previous work utilized the 2826.59 μm^2 . The optimized delay is 2.538 ns while previous delay is 434.4 ps. The power consumption is 450 μ W by proposed while 619.47 μ W in previous. Simulated results shows that proposed MEC-MED VLSI architecture gives significant improved results than conventional SEC-SED.

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